

1. Chebyshev Filter Design

SPECIFICATIONS:

Center frequency, f_0 :	2.4GHz
3 dB bandwidth, BW_{3dB} :	10%
Maximum ripple:	<0.1dB
Stopband bandwidth:	$2BW_{3dB}$
Stopband attenuation:	$\geq 36dB$
Source impedance, R_s :	50Ω
Load impedance, R_L :	50Ω
Relative permittivity of substrate, ϵ_r :	2.31
Thickness of the substrate, h :	31.5mil
Thickness of the conductor:	1.2mil

2. DESIGN STEPS

a. Find The Order Of Chebyshev Filter Needed

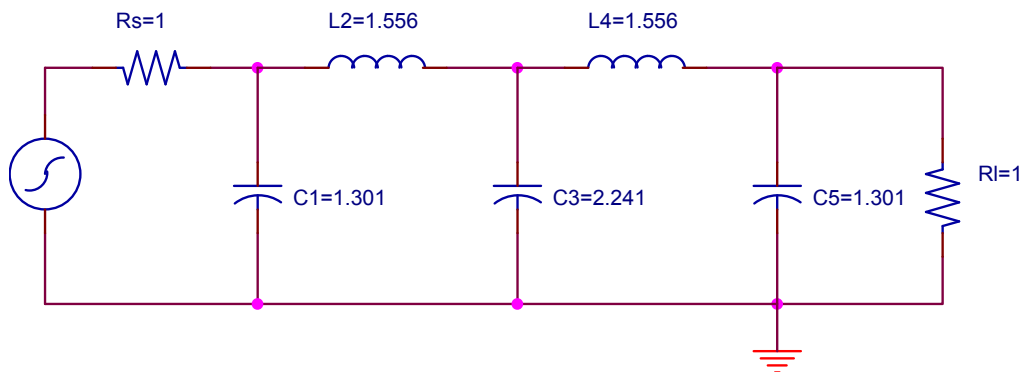
$$BW_{3dB} = 10\% \times f_0 = 240\text{MHz}$$

$$A_s = \frac{\text{StopbandBW}}{BW_{3dB}} = 2$$

Since the required stopband attenuation is at least 36dB, from the Attenuation characteristics curve for chebyshev filter with 0.1dB ripple, we find that the filter has to be at least 5th order.

The 5th order chebychev LPF prototype is presented in Figure 1.

Figure 1 5th order chebychev low pass filter prototype



b. Chebyshev BPF Using Discrete Components

The frequency-scaling factor is,

$$FSF = 2\pi BW = 2\pi \times 240\text{MHz}$$

And the impedance scaling factor is $Z=50$.

Therefore, we can find the actual component values of C_1 , L_2 , C_3 , L_4 and C_5 .

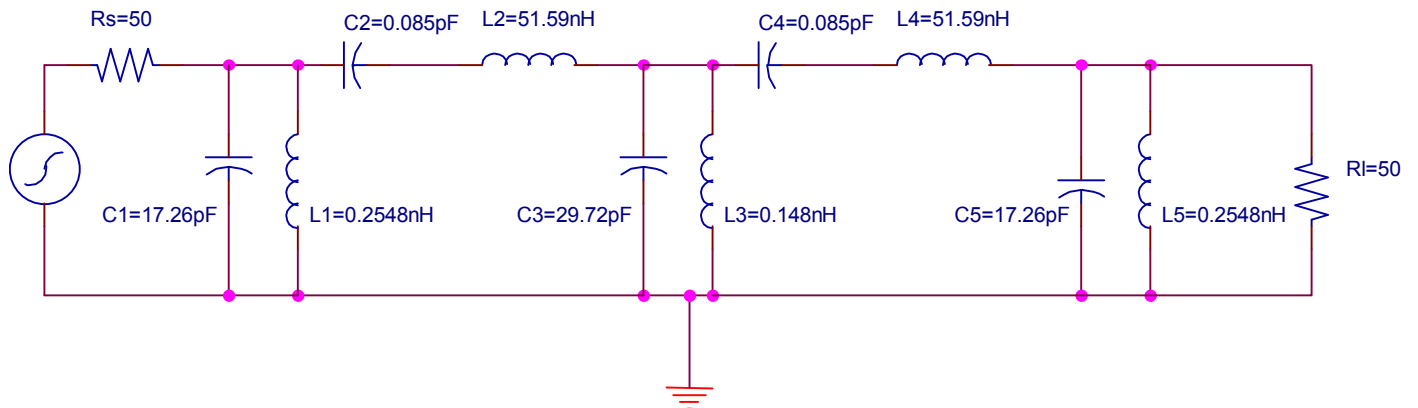
$$C_1 = C_5 = \frac{1.301}{FSF \times 50} = 17.26\text{pF}$$

$$L_2 = L_4 = \frac{1.556 \times 50}{FSF} = 51.59\text{nH}$$

$$C_3 = \frac{2.241}{FSF \times 50} = 29.72\text{pF}$$

To convert this LPF prototype to a BPF, we need to find the parallel L for each C and a series C for each L element, as shown in figure 2.

Figure 2 Chebyshev 5th order BPF using discrete components



Where,

$$L_1 = L_5 = \frac{1}{(2 \times \pi \times f_0)^2 \times C_1} = \frac{1}{(2 \times \pi \times f_0)^2 \times 17.26\text{pF}} = 0.2548\text{nH}$$

$$C_2 = C_4 = \frac{1}{(2 \times \pi \times f_0)^2 \times L_2} = \frac{1}{(2 \times \pi \times f_0)^2 \times 51.59\text{nH}} = 0.085\text{pF}$$

$$L_3 = \frac{1}{(2 \times \pi \times f_0)^2 \times C_3} = \frac{1}{(2 \times \pi \times f_0)^2 \times 29.72\text{pF}} = 0.148\text{nH}$$

c. Microstrip BPF Design

From the LFP prototype we have,

$$K_1 = K_5 = 1.301$$

$$K_2 = K_4 = 1.556$$

$$K_3 = 2.241$$

$$\Omega = BW_{3dB}/f_0 = 0.1$$

Therefore, we get,

$$J_{01}' = \left[\frac{\pi \times \Omega}{2 \times K_0 \times K_1} \right]^{0.5} = \left[\frac{\pi \times 0.1}{2 \times 1 \times 1.301} \right]^{0.5} = 0.347$$

$$J_{12}' = \frac{\pi \times \Omega}{2} \times \frac{1}{\sqrt{K_1 \times K_2}} = \frac{\pi \times 0.1}{2} \times \frac{1}{\sqrt{1.301 \times 1.556}} = 0.1104$$

$$J_{23}' = \frac{\pi \times \Omega}{2} \times \frac{1}{\sqrt{K_2 \times K_3}} = \frac{\pi \times 0.1}{2} \times \frac{1}{\sqrt{1.556 \times 2.241}} = 0.084$$

$$J_{34}' = \frac{\pi \times \Omega}{2} \times \frac{1}{\sqrt{K_3 \times K_4}} = \frac{\pi \times 0.1}{2} \times \frac{1}{\sqrt{2.241 \times 1.556}} = 0.084$$

$$J_{45}' = \frac{\pi \times \Omega}{2} \times \frac{1}{\sqrt{K_4 \times K_5}} = \frac{\pi \times 0.1}{2} \times \frac{1}{\sqrt{1.556 \times 1.301}} = 0.1104$$

$$J_{56}' = \left[\frac{\pi \times \Omega}{2 \times K_5 \times K_6} \right]^{0.5} = \left[\frac{\pi \times 0.1}{2 \times 1.301 \times 1} \right]^{0.5} = 0.347$$

Given, $\epsilon_r = 2.31$ and $h = 31.5$ mil.

Using ADS tools, the "LineCal" function, we can find that $W^{50} = 92$ mil.

From Figure 9.25 of the reference material, we have,

For $J_{01}' = J_{56}' = 0.347$,

$\text{Log}_{10}(S/h) = -0.85 \rightarrow S = 4.45$ mil

$\frac{W}{W^{50}} = 0.754 \rightarrow W = 69.368$ mil

$$\text{For } J_{12}' = J_{45}' = 0.1104,$$

$$\text{Log}_{10}(S/h) = 0.04 \rightarrow S = 34.54 \text{ mil}$$

$$\frac{W}{W^{50}} = 0.974 \rightarrow W = 89.608 \text{ mil}$$

$$\text{For } J_{23}' = J_{34}' = 0.084,$$

$$\text{Log}_{10}(S/h) = 0.16 \rightarrow S = 45.53 \text{ mil}$$

$$\frac{W}{W^{50}} = 0.982 \rightarrow W = 90.3 \text{ mil}$$

d. Simulation Using ADS Tools

i. Simulation Circuit And Results For The BPF Prototype

The BPF prototype circuit is shown in Figure 3.

The S-parameter simulation is done for a frequency range from 2.1GHz to 2.7GHz. The simulation results can be found in Figure 4.

Marker 1 and 2 show the -3dB frequencies. The 3dB BW is calculated to be 240MHz . The band separation of Marker 3 and 4 is 480MHz and the attenuations are -38dB and -44dB , respectively.

From the figure we also can see that the maximum passband ripple is equal to 0.1dB .

ii. Simulation Circuit And Results For Microstrip BPF

The microstrip BPF circuit is constructed as shown in Figure 5.

The value of the width and separation of every microstrip line calculated in section 3 were used at the beginning. The loss tangent was set to zero for simplicity. The circuit was simulated. The result showed a big ripple in passband and the center frequency was not at 2.4GHz .

Later on, some optimization was done to bring the center frequency back to 2.4GHz and to minimize the passband ripple. The length, width and separation of all the microstrip lines were set as tunable variables for optimization. The optimization goals were set to achieve 2.4GHz as the center frequency, 0.1dB passband ripple and to set 2.28GHz and 2.52GHz as the -3dB frequencies.

The simulation results (shown in Figure 6) were much satisfactory after the optimization.

The final circuit of the microstrip BPF is shown in Figure 7 and the layout is shown in Figure 8.

e. Testing Of Fabricated Microstrip BPF

The fabricated microstrip BPF was tested. The measured waveforms can be found in the appendix A.

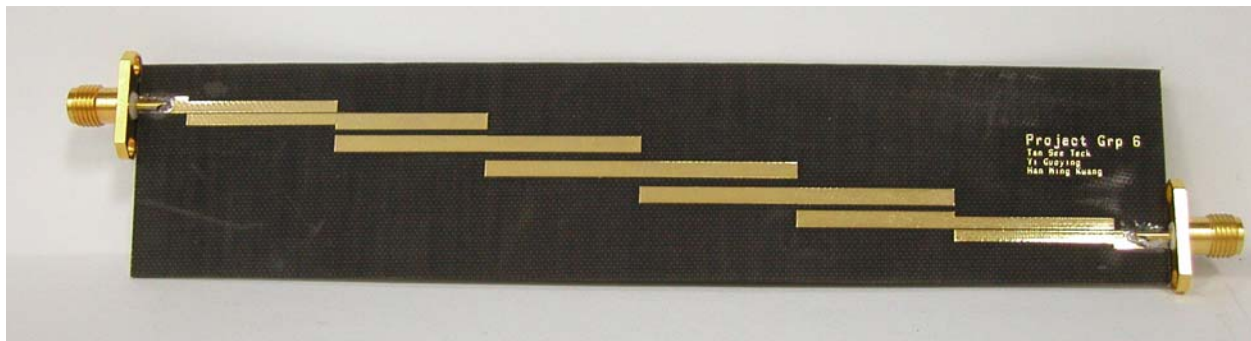
Table 1: Attenuation at different frequencies

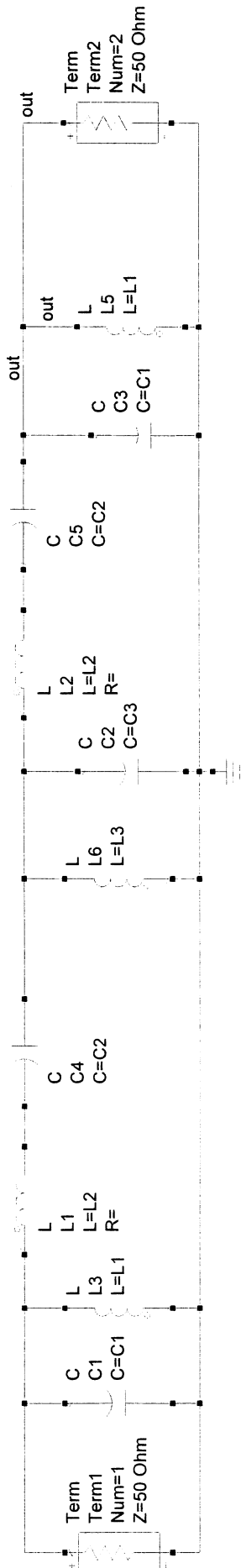
	2.4GHz	2.28GHz	2.52GHz	2.16GHz	2.64GHz
<i>Actual attenuation</i>	2.8262 dB	5.6721 dB	8.3403 dB	35.1 dB	37.859 dB
<i>Attenuation except the insertion loss</i>		2.846 dB	5.514 dB	32.27 dB	35.03 dB

From the above data, we can see that the actual filter performance is different from the design simulation.

In the design we have steeper roll off at the lower frequency side. But the above data shows that the high frequency side has more attenuation. Therefore, we suspect that the center frequency is shifted to the lower frequency side.

We also can see that the 3dB BW is narrower than the requirement. But the stop band didn't provide more than 36dB attenuation.





S-PARAMETERS
 S_Param
 AC1
 Start=2.1 GHz
 Stop=2.7 GHz
 Step=0.1 MHz

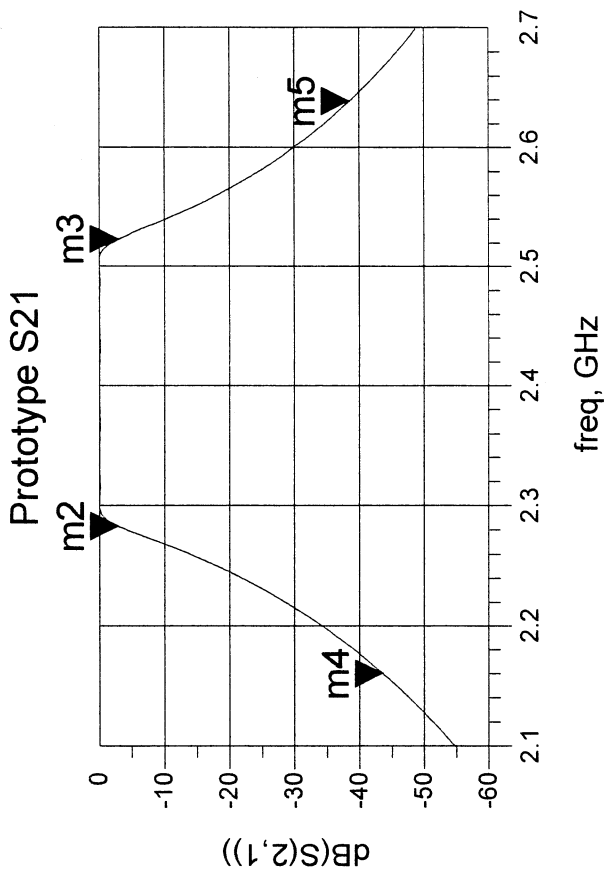
VAR
 VAR1
 Fo=2.4G
 Wo=2*pi*Fo
 Wo2=Wo*Wo
 BW=0.1*Fo
 Z=50

VAR
 VAR2
 g1=1.301
 g2=1.556
 g3=2.241

VAR
 VAR3
 C1=g1/(Z*2*pi*BW)
 L2=Z*g2/(2*pi*BW)
 C3=g3/(Z*2*pi*BW)

VAR
 VAR4
 L1=1/(Wo2*C1)
 C2=1/(Wo2*L2)
 L3=1/(Wo2*C3)

Figure 3: 5th order chebyshev BPF prototype

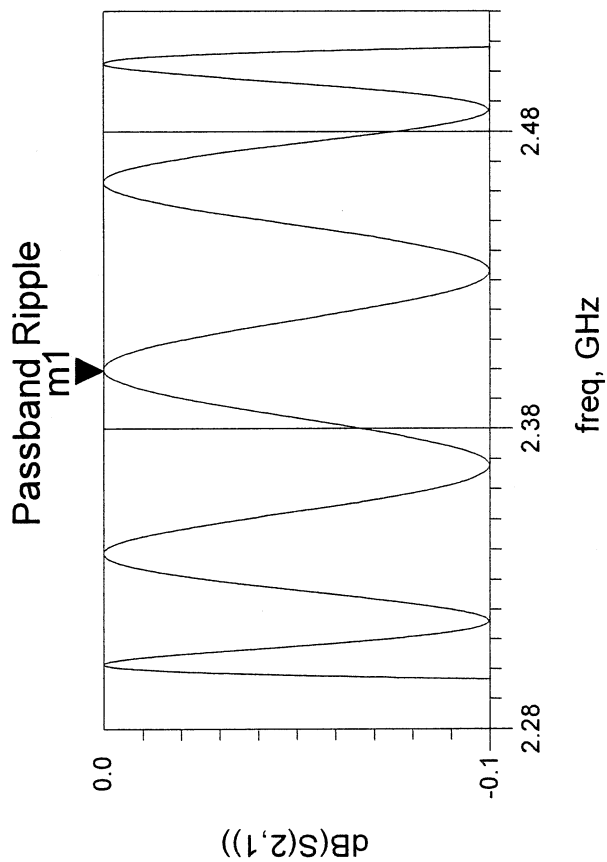


m2
freq=2.283GHz
dB(S(2,1))=-3.009

m3
freq=2.523GHz
dB(S(2,1))=-3.011

m4
freq=2.160GHz
dB(S(2,1))=-43.662

m5
freq=2.639GHz
dB(S(2,1))=-38.545



m1
freq=2.400GHz
dB(S(2,1))=-5.648E-5

Figure 4 : simulation result for prototype BPF

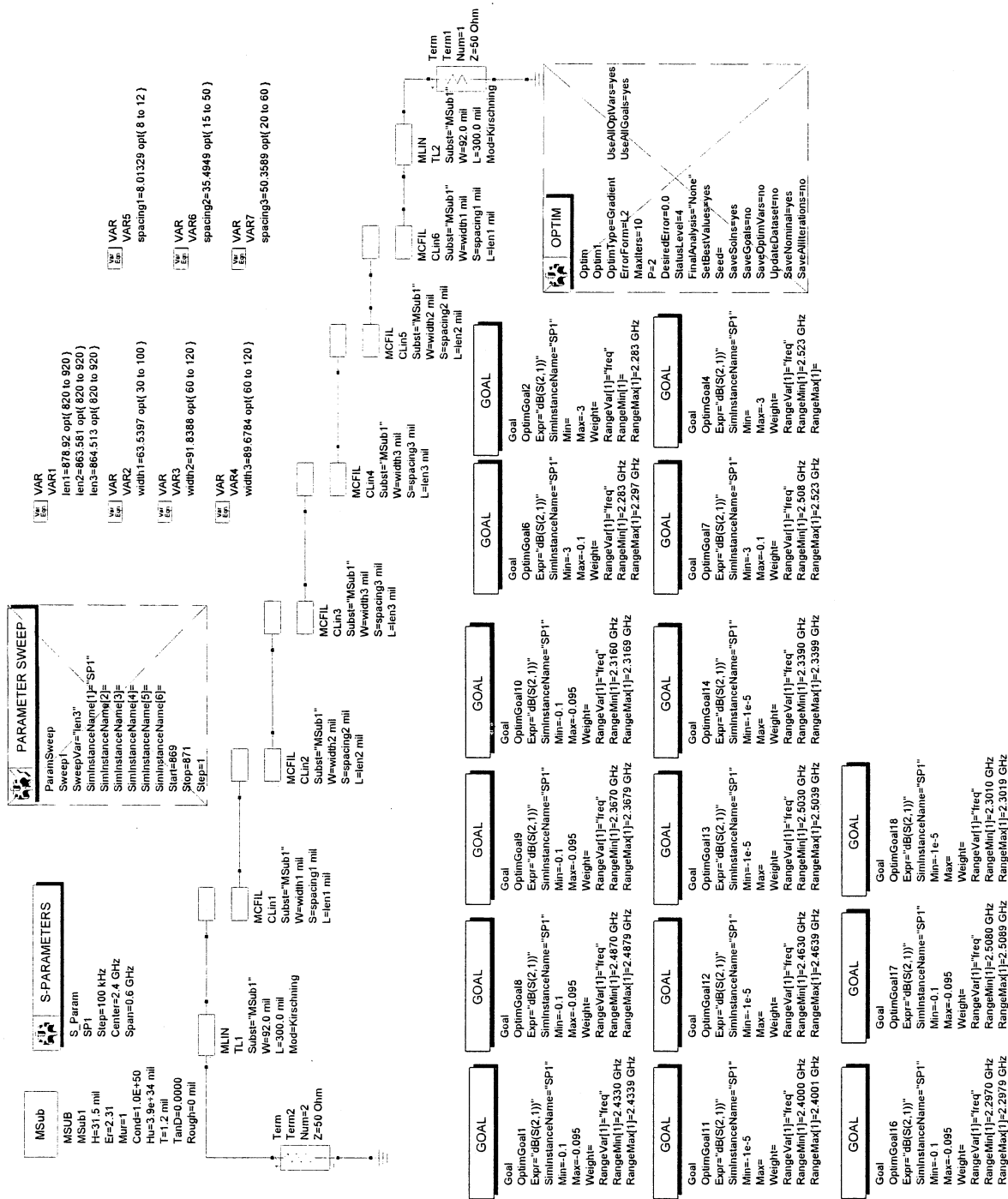
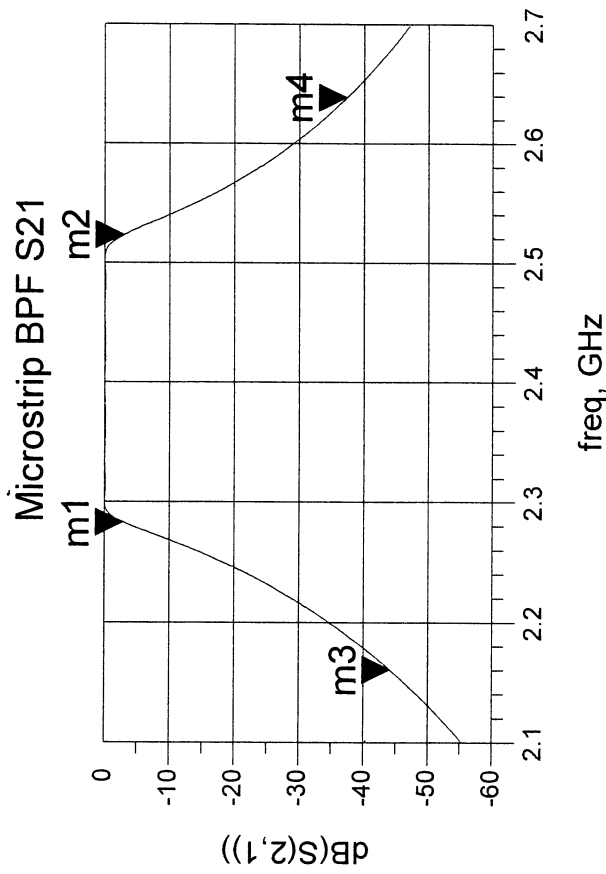


Figure 5. Microstrip BPF design

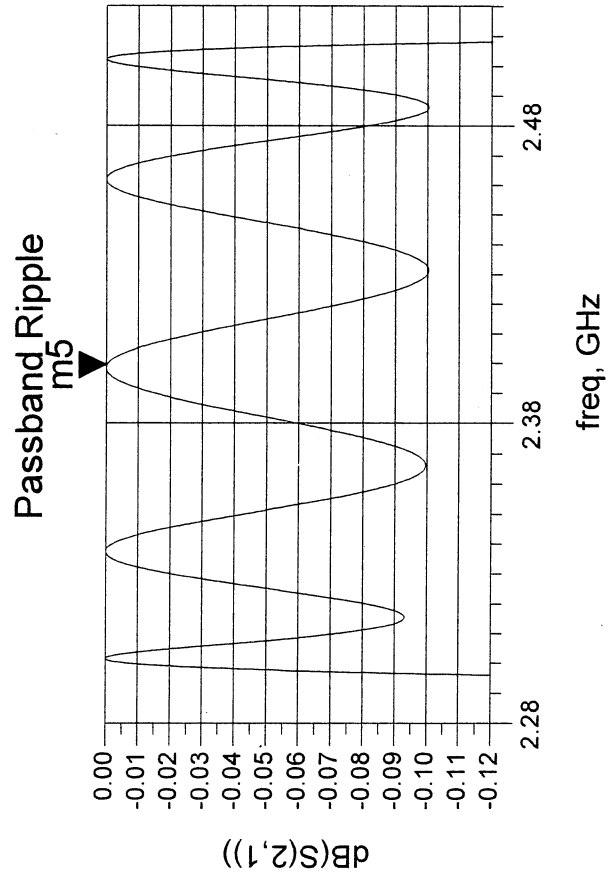


m1
freq=2.283GHz
dB(S(2,1))=-2.999

m2
freq=2.523GHz
dB(S(2,1))=-2.997

m3
freq=2.160GHz
dB(S(2,1))=-44.202

m4
freq=2.639GHz
dB(S(2,1))=-37.388



m5
freq=2.400GHz
dB(S(2,1))=-2.887E-4

Figure 6: Microstrip BPF simulation results

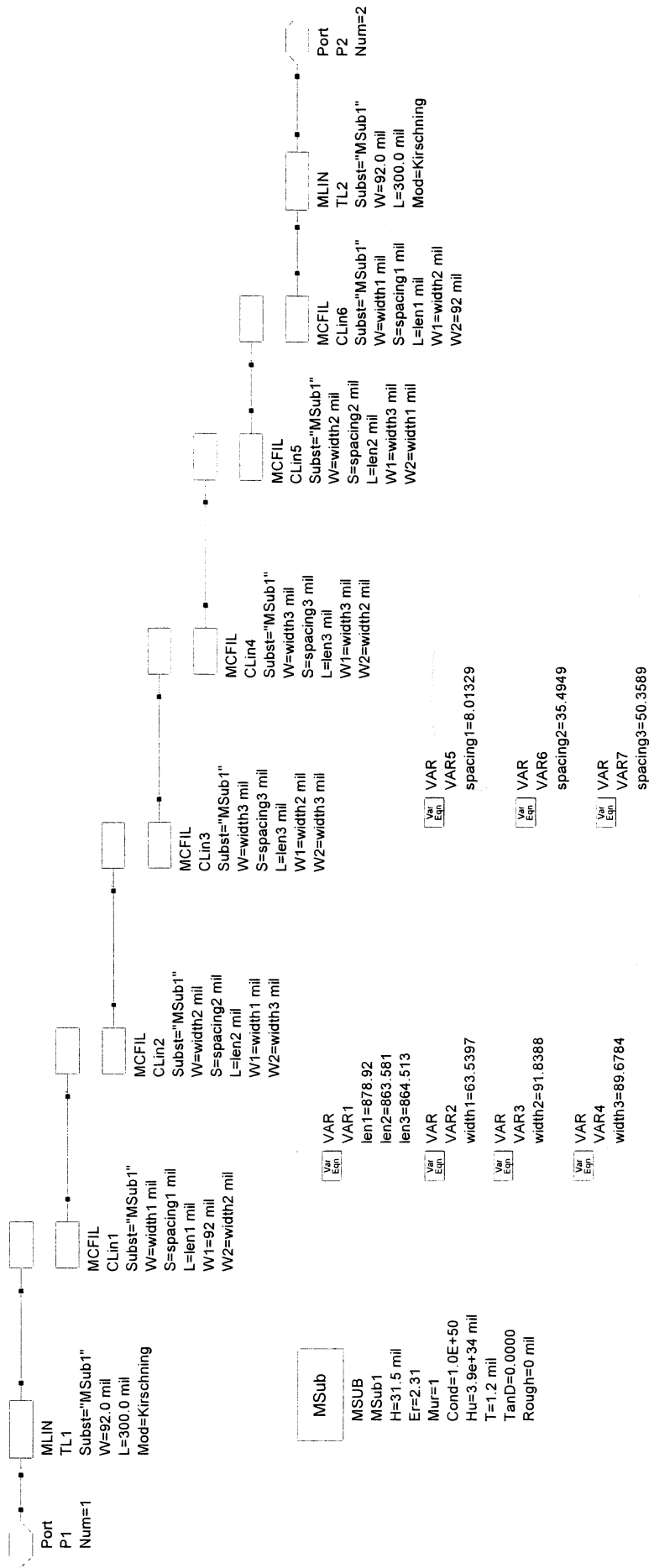


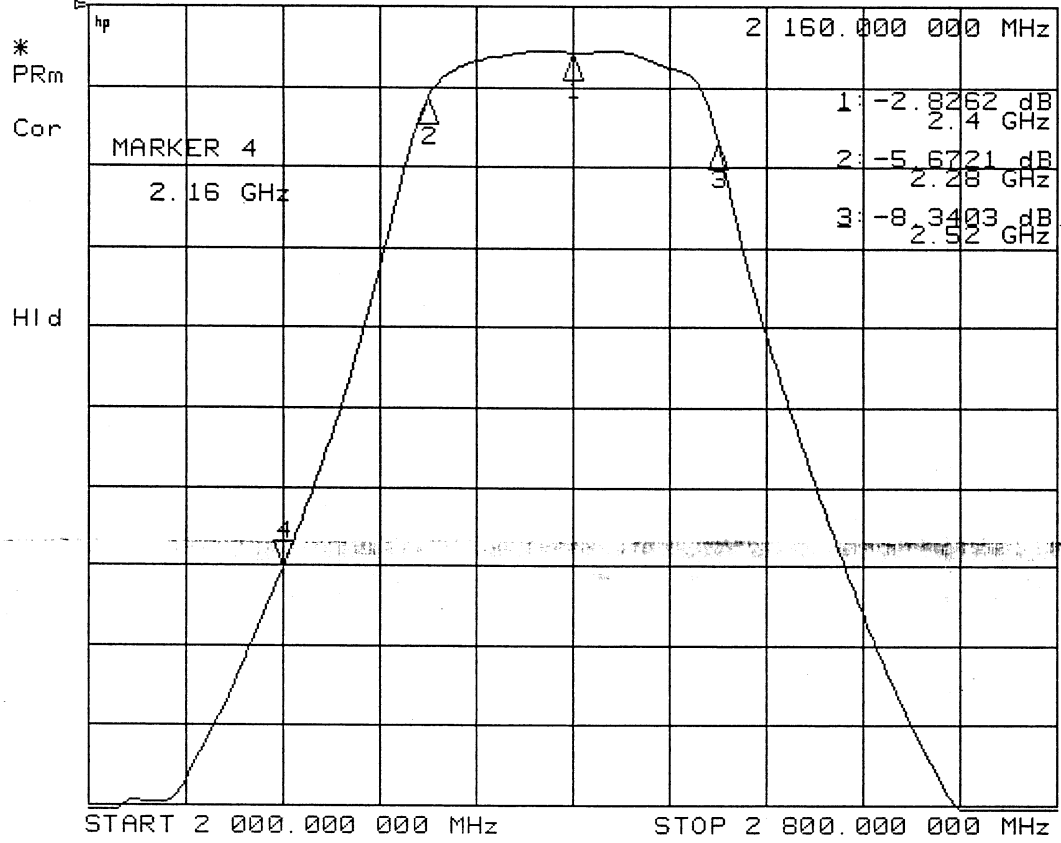
Figure 7. Final microstrip BPF schematic

APPENDIX A

TESTING RESULTS FOR MICROSTRIP CHEBYCHEV BAND PASS FILTER

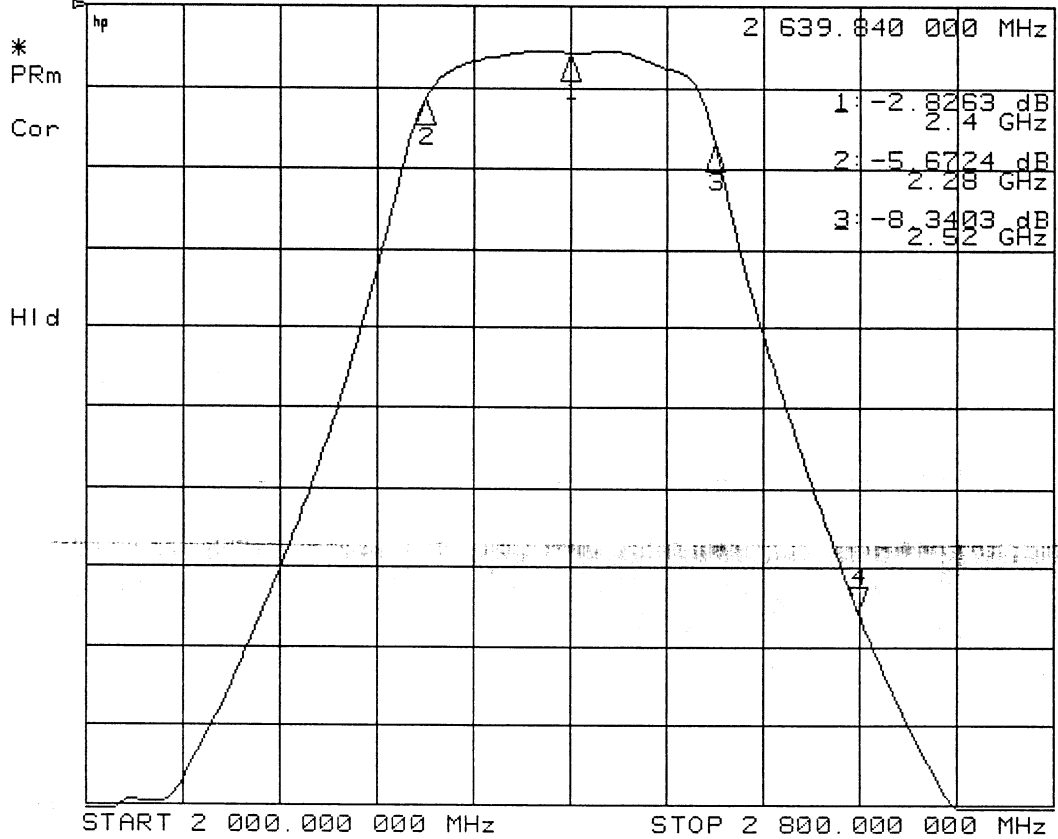
26 Mar 2003 19:32:23

CH1 S21 log MAG 5 dB/ REF 0 dB 4: -35.1 dB



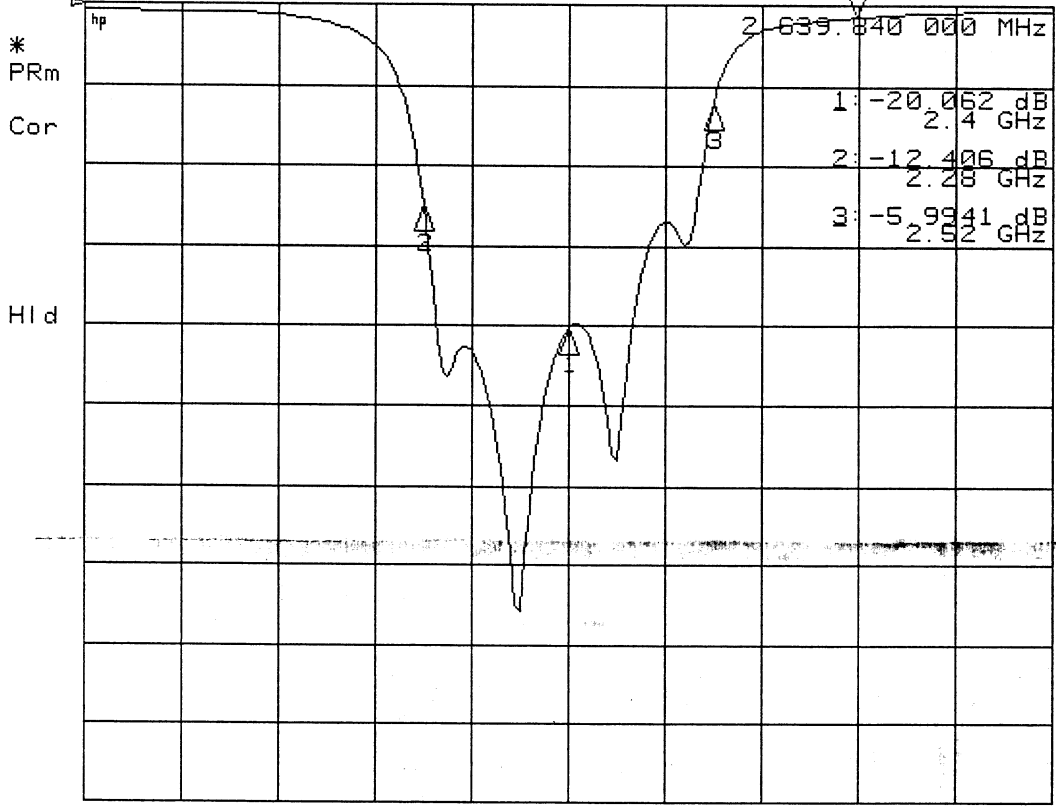
26 Mar 2003 19:27:06

CH1 S₂₁ log MAG 5 dB/ REF 0 dB 4: -37.859 dB



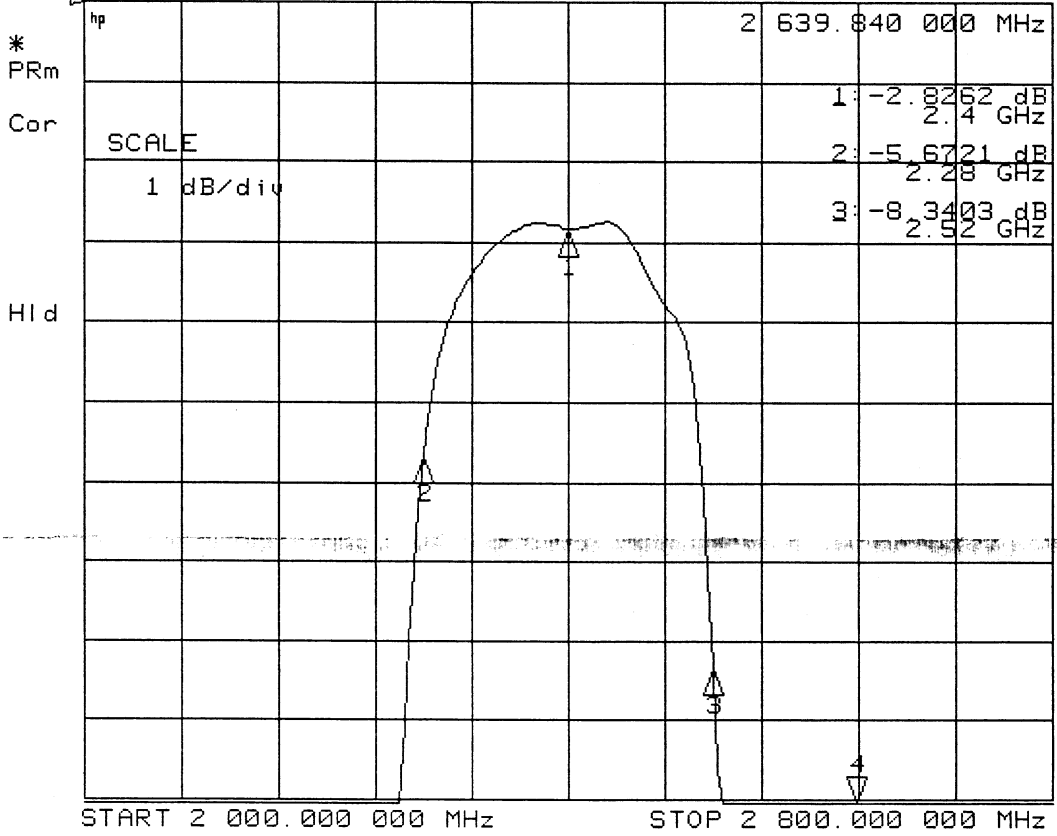
26 Mar 2003 19:27:57

CH1 S₁₁ log MAG 5 dB/ REF 0 dB 44 - .7007 dB



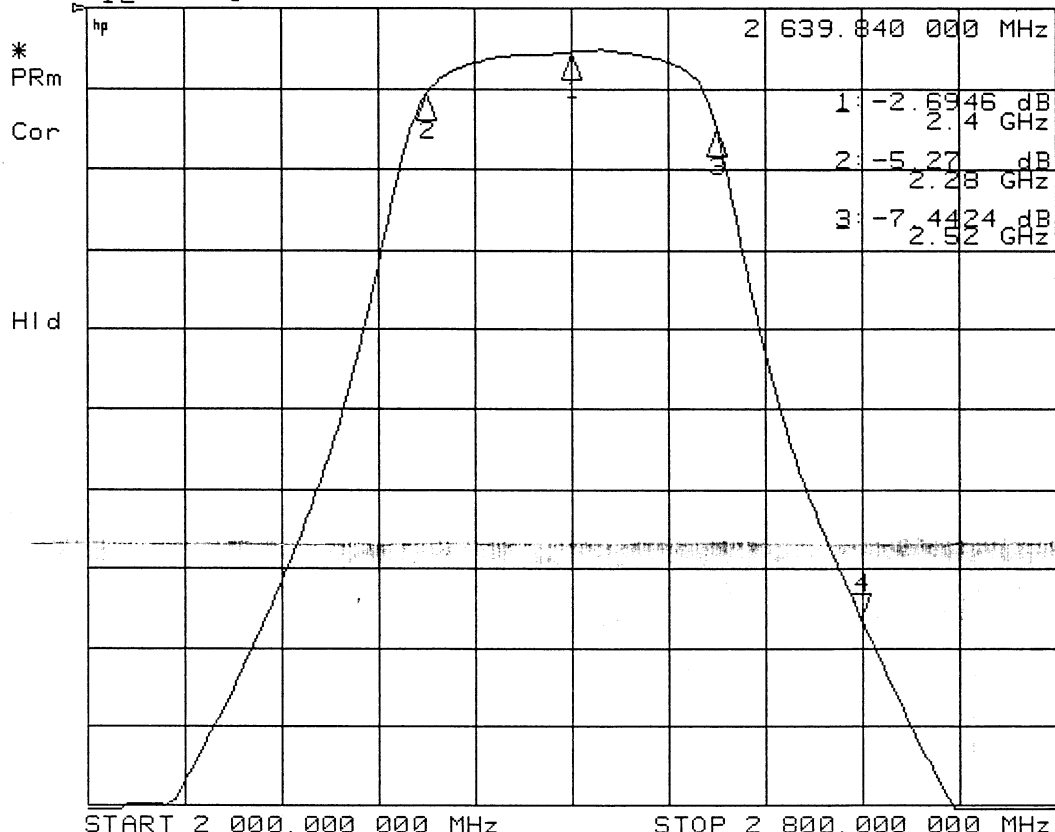
26 Mar 2003 19:30:06

CH1 S21 log MAG 1 dB/ REF 0 dB 4: -37.857 dB



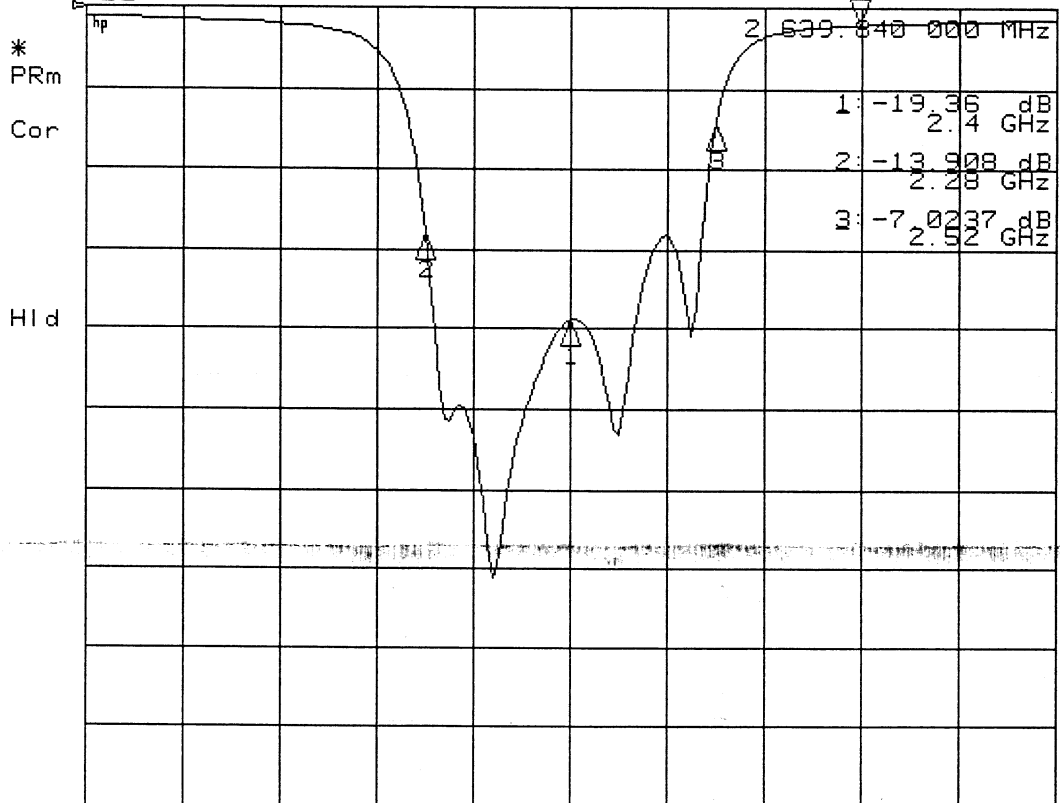
26 Mar 2003 19:28:12

CHI S12 log MAG 5 dB/ REF 0 dB 4: -38.225 dB



26 Mar 2003 19:28:38

CH1 S22 log MAG 5 dB/ REF 0 dB 44-1.0018 dB



START 2 000.000 000 MHz

STOP 2 800.000 000 MHz